A C/X Dual-band Wide-angle Reflective Polarization Rotation Metasurface

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Abstract. In this paper, a C/X dual-band wide-angle reflective polarization rotation metasurface (PRMS) with high rotation efficiency is proposed and realized. Aiming to miniaturize the size of the unit cell, a metallic flower-like shape ring is selected to extend the current path and the 45 degree slanting stitch along diagonal direction is used to form the asymmetric structure. The simulated results show that the proposed PRMS achieves polarization rotation at 4.61 GHz and 8.67 GHz with high efficiency, at which the linear polarization incident wave is converted into its orthogonal polarization after reflection. Furthermore, the high polarization rotation efficiency of the proposed PRMS is maintained under an oblique incident direction from 0° to 60°. To verify the simulated results, the proposed PRMS is fabricated and measured. The measured results are in good accordance with the simulated ones.

Keywords

Polarization rotation metasurface, dual band, wide angle

1. Introduction

As polarization is one of the most important characteristics of the electromagnetic (EM) waves, it is of great significance to be manipulated. Optical activity crystals and Faraday effects are traditionally used to realize the polarization rotation, whereas these methods usually require quite long propagation distance to get the phase accumulation resulting in bulky devices. Therefore, novel rotation devices with small thickness and light weight need to be developed.

Metasurface, as a two-dimensional artificial material, has received more and more interest for its novel electronic properties to realize cloaks [1], [2], absorbers [3–5], and polarization convertors [6–9]. With small thickness and light weight, the polarization rotation metasurface (PRMS) can realize polarization rotation in multi-band or broadband. Based on the way of controlling the polarization states, the PRMS can be divided into transmitted and reflective PRMS. For the transmitted PRMS, planar or 3D chiral metasurface have been proposed to control the polarization [10–12]. For the reflective PRMS, anisotropic metasurface with single- or multi-layer is employed to realize broadband PRMS [13–15]. Moreover, many structures are designed for better performance, such as V-shaped patches [14], U-shaped patches [16], and square patch loaded with metallic holes along the diagonal direction [13]. However, the polarization rotation efficiency of the PRMS under different incident directions is less discussed in previous articles. For wideband or multiband applications [16–22], the angle stability of the PRMS is still one of the research interests.

In this paper, a C/X dual-band wide-angle reflective PRMS with high polarization rotation efficiency is proposed. A flower-like shape ring with a 45 degree slanting stitch along diagonal direction is designed to form the anisotropy structure. The current path of the flower-like shape ring is larger, thus the size of the unit cell is smaller. The proposed PRMS can realize nearly 100% polarization rotation efficiency at 4.61 GHz and 8.67 GHz. Furthermore, the proposed PRMS can remain high polarization rotation efficiency under different incident directions from 0° to 60°. When the incident angle comes to 60°, the magnitude of PCR is still more than 79% at the resonant frequencies. The performance of the PRMS is analyzed by induced surface current distribution, physical theory and measurement results.

2. Design and Analysis

2.1 Design Principles

The diagram of the proposed units is shown in Fig. 1. The PRMS is etched on a substrate with period of 10.4 mm, thickness of 2.0 mm, dielectric constant of 4.4 and loss tangent of 0.001. A flower-like shape ring with a 45 degree slanting stitch along diagonal direction is on the top of the substrate and a full metallic plate is at the bottom.

As depicted in Fig. 1(b), the flower-like shape ring is a combination of four three-quarter metallic rings in a consecutive implementation. The current path of the flowerlike shape ring is larger, thus the size of the unit cell is



Fig. 1. The unit cell of the proposed PRMS: (a) Topological view; (b) Front view. The corresponding geometrical parameters are designed as: P = 10.4 mm, r1 = 3.05 mm, r2 = 1.75 mm, r3 = 2.55 mm, g1 = 0.2 mm and t = 2 mm.

smaller. The 45 degree slanting stitch along diagonal direction is used to form asymmetric structure, which can convert the linear polarization incident wave into its orthogonal polarization after reflection. Thus, the dualband polarization rotation is achieved through a miniaturized structure. The geometrical parameters are also given in Fig. 1.

2.2 Numerical Simulations

In order to verify the designed dual-band PRMS, numerical simulations were performed by using the commercial software HFSS. Periodic boundary condition and Floquet ports are used in HFSS. The metallic material is copper and the substrate is FR4. Owing to the anisotropy of the unit cell structure, the reflected wave generally consists of both co- and cross-polarized components. Resulting from a full metallic plate backed, the energy of the transmitted wave is zero, and the energy of the incident wave is all reflected ignoring the energy loss in the substrate.

For a y-polarized incident wave, the co- and crosspolarized components are defined as

$$\boldsymbol{r}_{yy} = \left| \boldsymbol{E}_{y} \right| / \left| \boldsymbol{E}_{y} \right|, \qquad (1)$$

$$r_{\rm yx} = \left| \boldsymbol{E}_{\rm rx} \right| / \left| \boldsymbol{E}_{\rm iy} \right|. \tag{2}$$

What's more, the polarization conversion ratio *PCR* is written as

$$PCR = r_{yx}^{2} / (r_{yx}^{2} + r_{xx}^{2}).$$
 (3)

To understand the polarization state of the reflective wave, the azimuth angle ψ and ellipticity angle κ are introduced and expressed as follows:

$$\tan(2\psi) = \frac{2r_{yx}\cos(\zeta)}{1 - r_{yy}^2},$$
 (4)

$$\sin(2\kappa) = \frac{2r_{\rm yx}\cos(\zeta)}{1+r_{\rm yx}^2},\qquad(5)$$

$$\zeta = Phase_{r_{yx}} - Phase_{r_{yy}}.$$
 (6)



Fig. 2. The reflection coefficient of the proposed PRMS: (a) the magnitude of r_{xy} and r_{yy} ; (b) PCR; (c) the phase of r_{xy} and r_{yy} ; (d) the azimuth angle ψ and ellipticity angle κ .

The simulated reflection results of the PRMS are shown in Fig. 2. Due to the less loss tangent of the dielectric substrate, the magnitude of r_{xy} is larger than 0.95 at 4.61 GHz and 8.67 GHz, at which the PCR is nearly 100%, shown in Fig. 2(a) and Fig. 2(b). It means that nearly all energy of y-polarized incident wave is converted into the x-polarized reflected one at these two key frequencies respectively. As depicted in Fig. 2(d), the azimuth angle ψ



Fig. 3. The angle stability of the proposed PRMS.

Cross polarization convertor	Resonance number	Angle stability	Period
[19]	3	Not mentioned	0.43λ ₀
[21]	2	45deg	0.30λ ₀
Our work	2	60deg	0.16λ ₀

Tab. 1. Performance comparison.

is nearly 0° at 4.61 GHz and 180° at 8.67 GHz. In addition, the ellipticity angle κ is approximately 2° at 4.61 GHz and -0.6° at 8.67 GHz, which indicates that the linear polarization incident wave can be converted into its orthogonal polarization after reflection.

The angle stability of the proposed PRMS is also investigated. The azimuth is defined by angle θ , indicating the azimuth between the incidence wave vector and PRMS surface vector. As the angle θ changing from 0° to 60°, the resonant frequencies and the PCR magnitude of the proposed PRMS are shown in Fig. 3. With the angle θ increasing, the first resonant frequency remains unchanged, and the second resonant frequency is slightly shifted to a higher frequencies is keeping unchanged or decreasing slightly, when the incident angle increases to 45°. What's more, when the incident angle comes to 60°, the magnitude of PCR is still coming to 86.5% at the first resonant frequency. The proposed PRMS possesses good angle stability.

Table 1 lists the performance comparison between three cross polarization convertors. Compared with the other two convertors, the proposed convertor has a smaller period and better angle stability at the same time. The overall length of the flower-like shape ring contains four continuously three-quarter circles, which is larger than previous literatures. Therefore, the size of the unit cell is smaller, only $0.16\lambda_0 \times 0.16\lambda_0$ (λ_0 represents free space wavelength at the minimal resonant frequency in the bandwidth). Moreover, the proposed PRMS is better in angle stability. The PCR at the second resonance frequency is near 80%, when the incident angle comes to 60°.

2.3 Physical and Theoretical Discussion

In order to get a physical insight into these plasmon frequencies, the surface current distribution on the flower-

like shape ring and the ground are depicted in Fig. 4. As demonstrated in Fig. 4, the overall direction of the surface current on the flower-like shape ring is antiparallel to those on the metallic ground presenting equivalent current loop. Then, the magnetic dipole is forming [23]. Moreover, the difference in the surface current directions indicates that the different magnetic responses are the results of the multi-order resonance modes.



Fig. 4. The surface current distribution of the PRMS: (a) Top surface current at 4.61 GHz; (b) Bottom surface current at 4.61 GHz; (c) Top surface current at 8.67 GHz; (d) Bottom surface current at 8.67 GHz.

To understand the response of the proposed PRMS to the y-polarized incident EM wave theoretically, we decompose the y-polarized incident wave into two orthogonal axis u and v, which are along the 45 degree direction with respect to axis x and axis y shown in Fig. 5. Hence, the ypolarized incident wave can be written as (7):

$$\boldsymbol{E}_{iy} = \left| \boldsymbol{E}_{iy} \right| \boldsymbol{e}_{y} = E_{iu} \boldsymbol{e}_{u} + E_{iv} \boldsymbol{e}_{v} .$$
(7)

Meanwhile, the reflective wave can be defined as (8):

$$\boldsymbol{E}_{\rm ry} = \left| \boldsymbol{E}_{\rm iy} \right| \boldsymbol{e}_{\rm y} = r_{\rm u} E_{\rm iu} \boldsymbol{e}_{\rm u} + r_{\rm v} E_{\rm iv} \boldsymbol{e}_{\rm v} \,. \tag{8}$$

Moreover, $\Delta \Phi$ is used to define the phase difference between r_u and r_v , and then the relationship between r_u and r_v can be written as (9):

$$r_{\rm v} = r_{\rm u} \exp\left(-j\Delta\Phi\right). \tag{9}$$

Thus, the reflective wave E_r can be modified by inserting (9) into (8):

$$\boldsymbol{E}_{\mathrm{r}} = r_{\mathrm{u}} E_{\mathrm{iu}} \boldsymbol{e}_{\mathrm{u}} + r_{\mathrm{u}} \exp\left(-j \varDelta \boldsymbol{\Phi}\right) E_{\mathrm{iv}} \boldsymbol{e}_{\mathrm{v}}.$$
 (10)

Due to the less tangent loss substrate and the anisotropy of the unit cell, the magnitude of r_u and r_v would almost be equal to 1, demonstrated in Fig. 6(a). When it comes to the resonant frequencies, the $\Delta \Phi$ is nearly 180°, depicted in Fig. 6(b). By again looking to (10) and setting $r_u = r_v \approx 1$, and $\Delta \Phi \approx 180^\circ$, Equation (10) is equal to

$$\boldsymbol{E}_{\mathrm{r}} = r_{\mathrm{u}} E_{\mathrm{iu}} \boldsymbol{e}_{\mathrm{u}} + r_{\mathrm{u}} \exp(-j\Delta \boldsymbol{\Phi}) E_{\mathrm{iv}} \boldsymbol{e}_{\mathrm{v}}$$
$$= E_{\mathrm{iu}} \boldsymbol{e}_{\mathrm{u}} + \exp(-j180^{\circ}) E_{\mathrm{iv}} \boldsymbol{e}_{\mathrm{v}} \qquad (11)$$
$$= E_{\mathrm{iu}} \boldsymbol{e}_{\mathrm{u}} - E_{\mathrm{iv}} \boldsymbol{e}_{\mathrm{v}} = E_{\mathrm{v}} \boldsymbol{e}_{\mathrm{x}}.$$

Thus, a dual-band polarization rotation metasurface is achieved.



Fig. 5. Intuitive image of y-to-x polarization conversion.

3. Fabrication and Measurement

To verify our simulation results, the proposed PRMS is fabricated using a combination of 18×18 unit cells with an area of 187.2 mm × 187.2 mm, shown in Fig. 7(a). Agilent N5230C vector network analyzer, connected with two standard linearly-polarized horn antennas, is used to



Fig. 6. The reflection coefficient of r_{uu} and r_{vv} : (a) The magnitude; (b) The phase and phase difference.





Fig. 7. The photograph of the proposed PRMS: (a) The fabrication; (b) The experimental setup.

measure the reflection coefficient of the proposed PRMS in an EM anechoic chamber, depicted in Fig. 7(b). The sample was placed at a distance of 150 cm away from the horn antennas. Moreover, the undesirable signals are eliminated



Fig. 8. The measured PCR of the proposed PRMS.

by the time-domain gating. To measure the reflected copolarization and cross-polarization wave, the receiving antenna needs to be rotated 0° and 90° .

The measured results of the reflection coefficient and PCR are shown in Fig. 8. The measurement results show that the proposed PRMS can convert the linearly-polarized wave to its orthogonal component at 4.60 GHz and 8.50 GHz with high polarization rotation efficiency. The measurement results are in good agreement with the simulation results. The small difference between measurement results and simulation ones might be due to: 1. Some manufacturing error exists in the fabrication; 2. The dimension of PRMS is infinite in the simulation, while the dimension of the fabricated PRMS is finite resulting in the edge diffraction; 3. The incident wave is slightly oblique in the measurements, while the incident wave is under normal direction in the simulation.

4. Conclusion

In this paper, a C/X dual-band wide-angle reflective PRMS is demonstrated, with numerical simulation, theoretical analysis and experimental measurement. The proposed PRMS realizes nearly 100% polarization rotation efficiency at 4.61 GHz and 8.67 GHz, at which the linear polarization incident wave can be converted into its orthogonal polarization after reflection. Besides, the PRMS can remain high polarization rotation efficiency under an oblique incident direction from 0° to 60°. The measurement results are in good agreement with the simulation ones. The proposed PRMS has great potential to realize dual-band polarization rotation, radar section cross reduction, and so on.

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